

**NATIONAL EXAMS December 2011**  
**07-Elec-B2 Advanced Control Systems**

3 hours duration

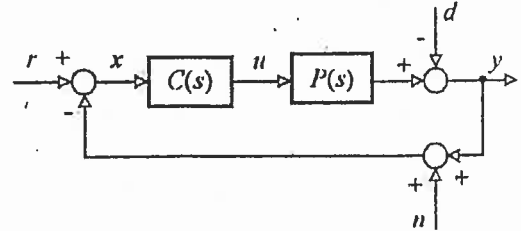
NOTES:

1. If a doubt exists as to the interpretation of any question, the candidate is urged to submit with the answer paper, a clear statement of any assumptions made.
2.                      a Casio or Sharp approved calculator.
3. This is a closed-book examination. Tables of Laplace and z-transforms are attached as page 4 and page 5.
4. Any four questions constitute a complete paper. Only the first four questions as they appear in your answer book will be marked.
5. All questions are of equal value.

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1. Consider the control system below with,  $P(s) = \frac{30}{(s+3)(s+10)}$  and  $C(s) = \frac{K_1 + sK_2}{s}$ .

- Let  $K_2 = 0$ . Determine a value of  $K_1$  say  $K_1 = K_{1M}$  such that the closed loop system exhibits sustained oscillation. Determine the oscillation frequency.
- For  $K_1 = K_{1M}$  determine a value for  $K_2$  such that the phase margin exceeds 45 degrees.
- Determine the steady state value of  $y(t)$  when  $d(t) = a$  ramp of slope 3, and  $r(t) = a$  unit step, and  $n(t) = 0$ .



2. Consider the system,

$$\dot{x}(t) = Ax(t) + Bu(t), \quad A = \begin{pmatrix} 0 & 0 & -3 \\ 0 & 0 & 1-3\alpha \\ 0 & 1 & 0 \end{pmatrix}, \quad B = \begin{pmatrix} 1 \\ \alpha \\ 0 \end{pmatrix}$$

$$y(t) = Cx(t) + Du(t), \quad C = (1 \ 1 \ 0), \quad D = 0$$

- Determine the conditions for controllability and observability.
- For  $\alpha = 2$ , design a controller of the form  $u(t) = Lr(t) - Kx(t)$ , i.e., determine  $L$  and  $K$  such that the closed loop poles are  $s = -2 + j$ ,  $s = -2 - j$ ,  $s = -4$ , and the steady state tracking error,  $e = r - y$ , is zero when  $r(t)$  is a unit step.

3. The discrete time model for a system has the form,  $y(z) = P(z)u(z)$  where  $P(z) = \frac{4z^{-2}}{1-z^{-1}}$ .

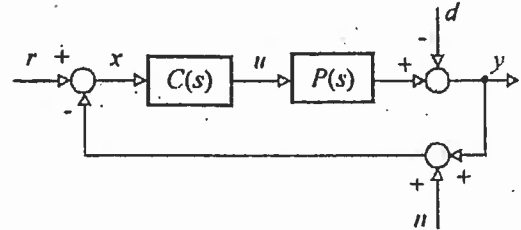
- Consider a discrete proportional feedback controller,  $K$ . Determine the range of  $K$  that ensures closed loop stability.
- Draw the associated root locus.
- Assuming  $y(k)$  is sampled uniformly from the output of a continuous linear time invariant system and the sample  $u(k)$  is applied uniformly to a ZOH that drives the input of the continuous time system, determine a transfer function for the continuous time system assuming the sample period is  $h = 1$  s.

4. Several experiments are conducted on an unknown system,  $P(s)$  :

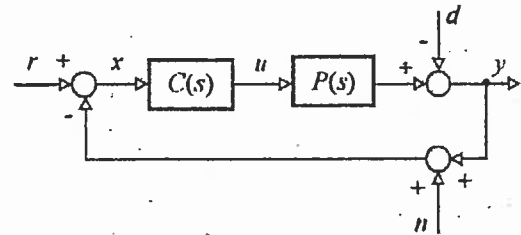
- When a step of magnitude 5 is applied to the input, the steady state output is 20 and the maximum overshoot is 30.
- When a sinusoid of frequency 8 rad/sec is applied, the phase lag at the output is  $90^\circ$ .

- Assume the system,  $P(s)$ , is second order system and has no finite zeros. Find the parameters of the second order model.
- Justify whether or not it is feasible to reduce the percentage overshoot by using a proportional feedback controller.

5. Consider the feedback system below with  $P(s) = \frac{4(s+3)}{s^2 + 0.2s + 2}$ .
- Determine a feedback controller,  $C(s)$ , such that the closed loop transfer function relating  $r$  to  $y$  is given by  $\frac{32}{8s^2 + 3s + 32}$ . Note:  $C(s)$  must be proper, i.e., the degree of the numerator must be greater than or equal to the degree of the denominator.
  - Determine the gain and phase margin of the feedback design.
  - Determine the steady state value of  $u$  when  $d$  is a unit step,  $r = 0$ , and  $n = 0$ .



6. Consider the feedback system below with,  $C(s) = \frac{K}{s}$ ,  $P(s) = \frac{20e^{-s}}{s+10}$ .
- Design  $C(s)$  such that the gain margin is 6dB.
  - Determine the phase margin.
  - The system is operating with constant inputs,  $n(t) = 0$ ,  $d(t) = d_0$ , and  $r(t) = r_0$ . Determine the steady state tracking error,  $e(t) = r(t) - y(t)$ , as a function of  $K$ .



Inverse Laplace Transforms	
$F(s)$	$f(t)$
$\frac{A}{s+\alpha}$	$Ae^{-\alpha t}$
$\frac{C+jD}{s+\alpha+j\beta} + \frac{C-jD}{s+\alpha-j\beta}$	$2e^{-\alpha t} (C \cos \beta t + D \sin \beta t)$
$\frac{A}{(s+\alpha)^{n+1}}$	$\frac{At^n e^{-\alpha t}}{n!}$
$\frac{C+jD}{(s+\alpha+j\beta)^{n+1}} + \frac{C-jD}{(s+\alpha-j\beta)^{n+1}}$	$\frac{2t^n e^{-\alpha t}}{n!} (C \cos \beta t + D \sin \beta t)$

Inverse z-Transforms	
$F(z)$	$f(nT)$
$\frac{Kz}{z-a}$	$Ka^n$
$\frac{(C+jD)z}{z-re^{j\varphi}} + \frac{(C-jD)z}{z-re^{-j\varphi}}$	$2r^n (C \cos n\varphi - D \sin n\varphi)$
$\frac{Kz}{(z-a)^r}, r=2,3,\dots$	$\frac{Kn(n-1)\dots(n-r+2)}{(r-1)!a^{r-1}} a^n$

<b>Table of Laplace and z-Transforms</b> ( <i>h</i> denotes the sample period)		
$f(t)$	$F(s)$	$F(z)$
unit impulse	1	1
unit step	$\frac{1}{s}$	$\frac{z}{z-1}$
$e^{-\alpha t}$	$\frac{1}{s+\alpha}$	$\frac{z}{z-e^{-\alpha h}}$
$t$	$\frac{1}{s^2}$	$\frac{hz}{(z-1)^2}$
$\cos \beta t$	$\frac{s}{s^2 + \beta^2}$	$\frac{z(z - \cos \beta h)}{z^2 - 2z \cos \beta h + 1}$
$\sin \beta t$	$\frac{\beta}{s^2 + \beta^2}$	$\frac{z \sin \beta h}{z^2 - 2z \cos \beta h + 1}$
$e^{-\alpha t} \cos \beta t$	$\frac{s+\alpha}{(s+\alpha)^2 + \beta^2}$	$\frac{z(z - e^{-\alpha h} \cos \beta h)}{z^2 - 2ze^{-\alpha h} \cos \beta h + e^{-2\alpha h}}$
$e^{-\alpha t} \sin \beta t$	$\frac{\beta}{(s+\alpha)^2 + \beta^2}$	$\frac{ze^{-\alpha h} \sin \beta h}{z^2 - 2ze^{-\alpha h} \cos \beta h + e^{-2\alpha h}}$
$t f(t)$	$-\frac{dF(s)}{ds}$	$-zh \frac{dF(z)}{dz}$
$e^{-\alpha t} f(t)$	$F(s + \alpha)$	$F(ze^{\alpha h})$